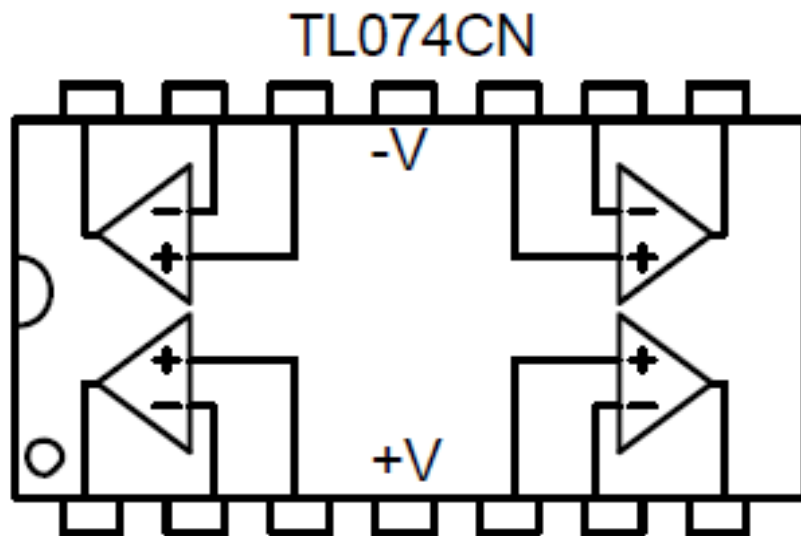
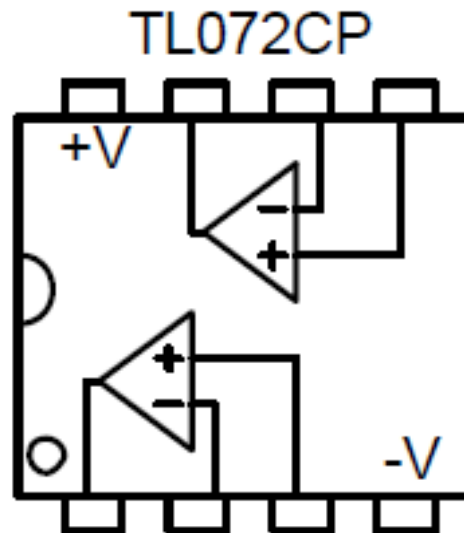
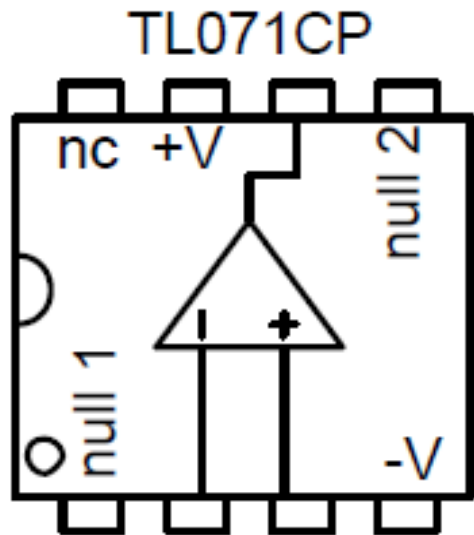


Using TL07X Op Amps in Analog Synthesizers

Presented by Ray Wilson of MFOS

- Basic TL07X Information
- Inverting gain
- Non-inverting gain
- Input/Output Coupling
- AC vs. DC Input Mixer
- Precision Full Wave Rectifier
- Fuzz Tone
- Comparators
- Comparator Hysteresis
- Pulse Width Modulation
- Window Comparator
- Capacitors to the Rescue
- Integrator
- Battery Power
- Active Lowpass, Highpass, and Bandpass Filters
- Online Filter Calculators You Should Know About

Basic TL07X Information



Read the data sheet for full details.

Maximum Supply V: +/-18V

Minimum Supply V: +/-3V

Maximum Input V: +/-15V

Unity GBW = 3MHz.

Current per amp \approx 2mA.

High input impedance (10^{12} ohm)

Output voltage swing with supply

+/-9V \approx +/-7V

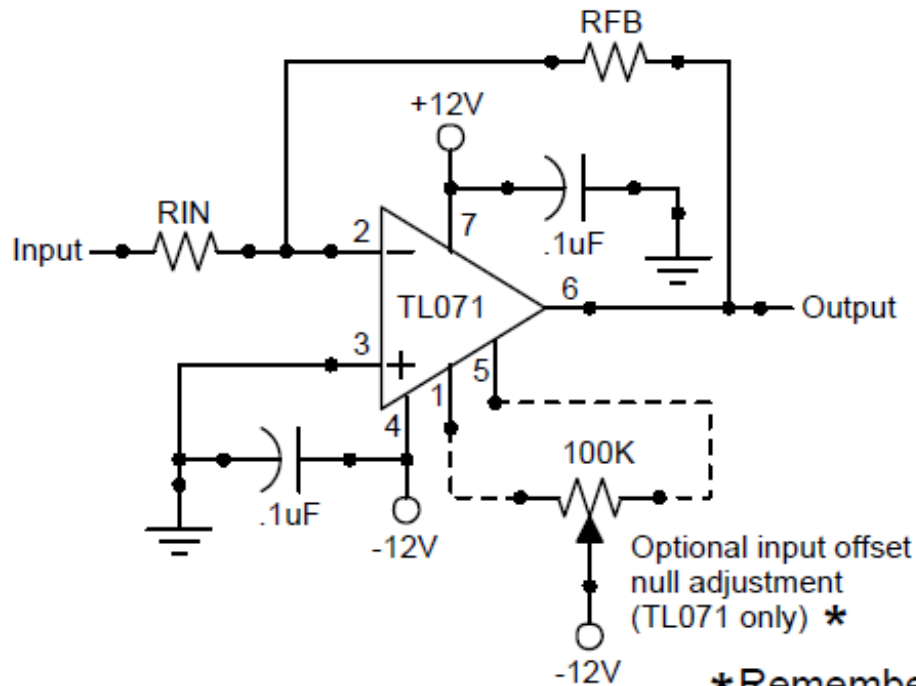
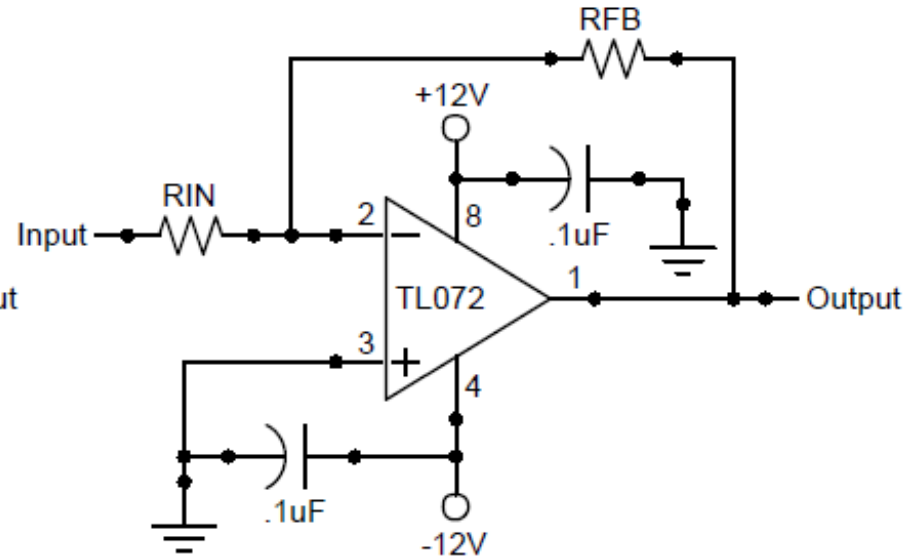
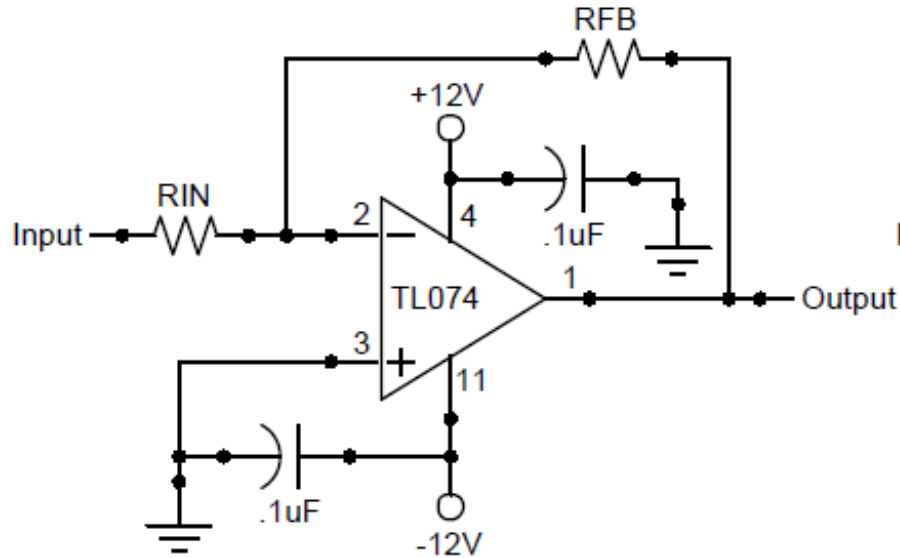
+/-12V \approx +/-10V

+/-15V \approx +/-13.5V

Keep output load resistance above 2K
for best audio results.

ALWAYS bypass the package with
0.1uF ceramic caps close to the power
pins.

Inverting Gain



Optional input offset null adjustment (TL071 only) *

*Remember - Input offset voltage gets amplified too!

Inverting Gain Formula

$$\text{Voltage Gain} = -1 \left(\frac{R_{FB}}{R_{IN}} \right)$$

Examples:

$R_{IN} = 10K, R_{FB} = 100K, \text{Gain} = -10$

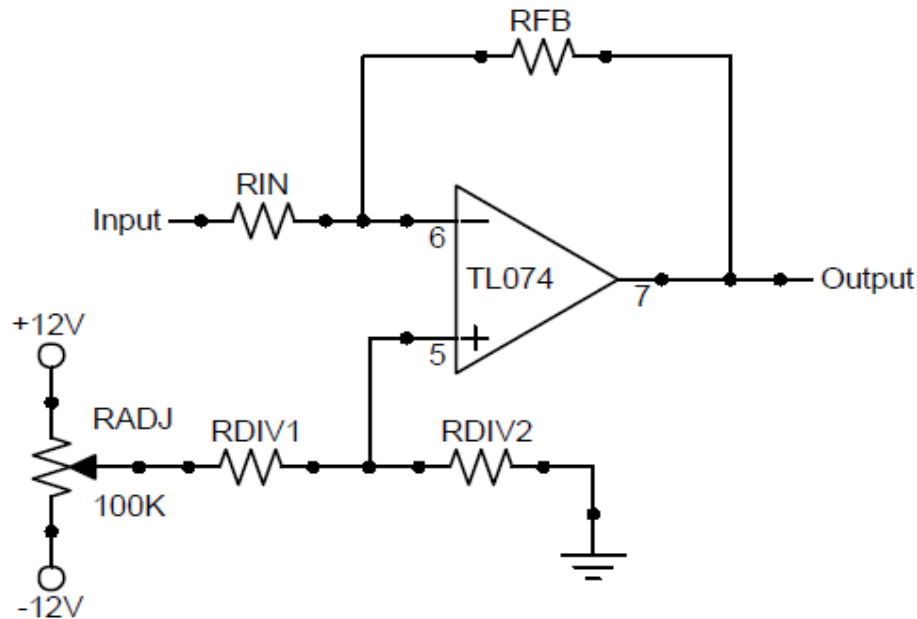
$R_{IN} = 20K, R_{FB} = 20K, \text{Gain} = -1$

$R_{IN} = 100K, R_{FB} = 2K, \text{Gain} = -.02$

Offset Adjust or Biasing an Inverting Op Amp

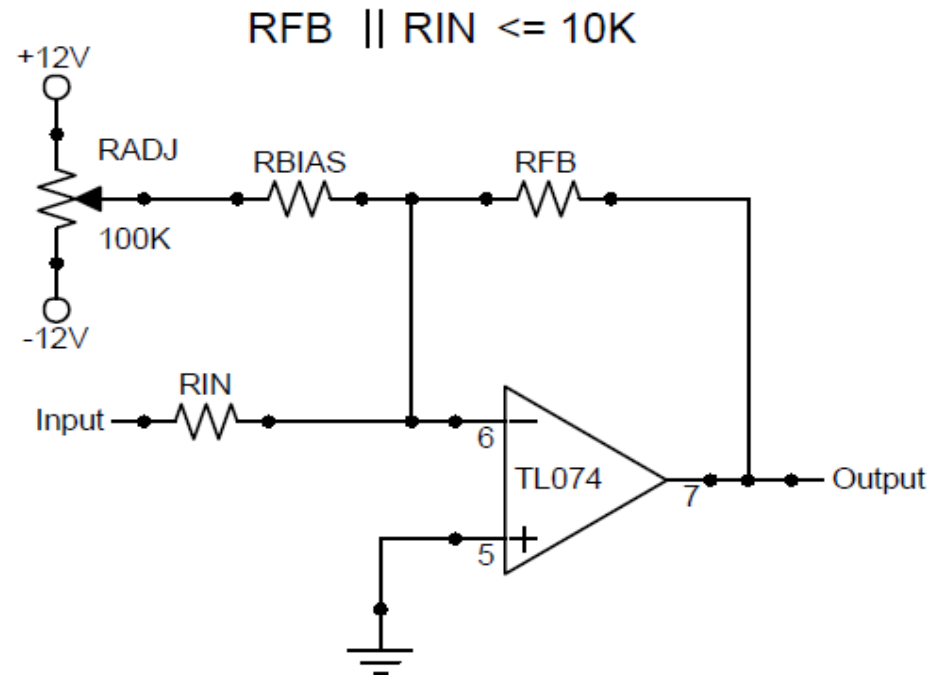
For more information see: National Semiconductor LINEAR APPLICATIONS HANDBOOK
Linear Brief 9 - "Universal Balancing Techniques", Aug 1969 by Robert C. Dobkin

$$\text{Range} = \pm V \left(\frac{R_{DIV2}}{R_{DIV1}} \right) \times \text{Gain}$$



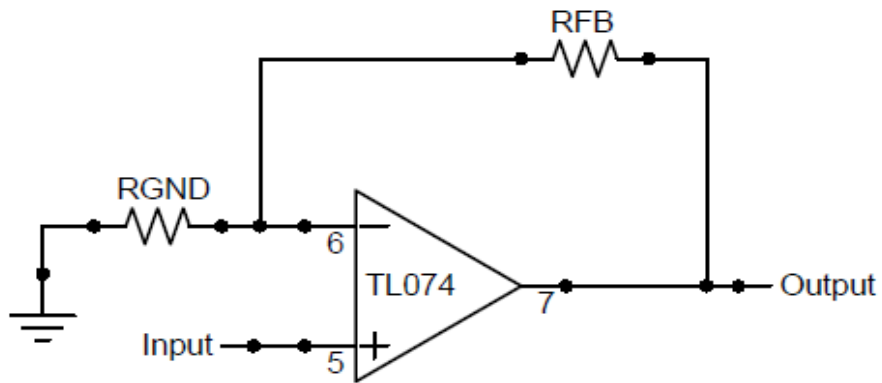
The gain supplied by the op amp's inverting feedback configuration will multiply the bias voltage applied to the non-inverting input so for high gain scenarios make RDIV1 high in relation to RDIV2.

$$\text{Range} = \pm V \left(\frac{R_{FB} \parallel R_{IN}}{R_{BIAS}} \right)$$

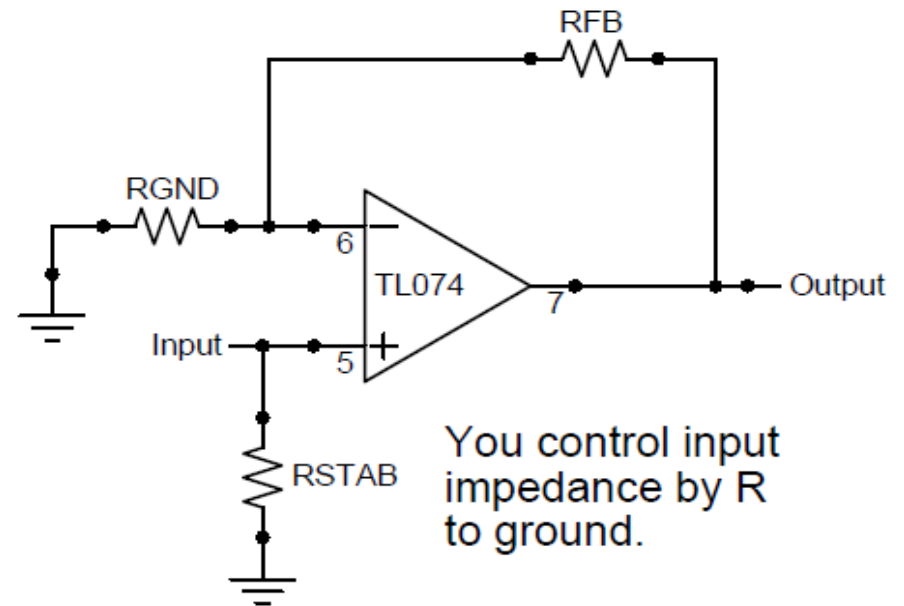


RBIAS should be selected based on the range of adjustment needed. Higher values of RBIAS will result in a smaller range of adjustment and vice versa.

Non-Inverting Gain



Highest impedance condition with no resistor to ground.



You control input impedance by R to ground.

Non-inverting Gain Formula

$$\text{Voltage Gain} = \left(\frac{R_{FB}}{R_{GND}} \right) + 1$$

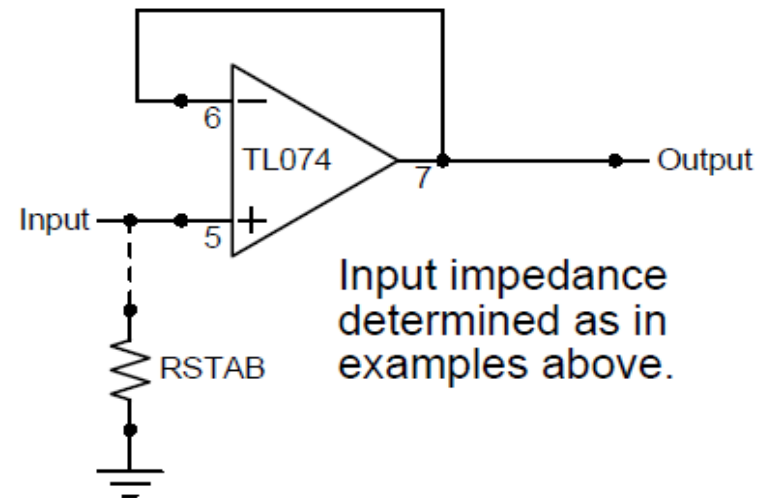
Examples:

$R_{GND} = 10K, R_{FB} = 100K, \text{Gain} = 11$

$R_{GND} = 20K, R_{FB} = 20K, \text{Gain} = 2$

$R_{GND} = 100K, R_{FB} = 2K, \text{Gain} = 1.02$

Unity Gain Follower



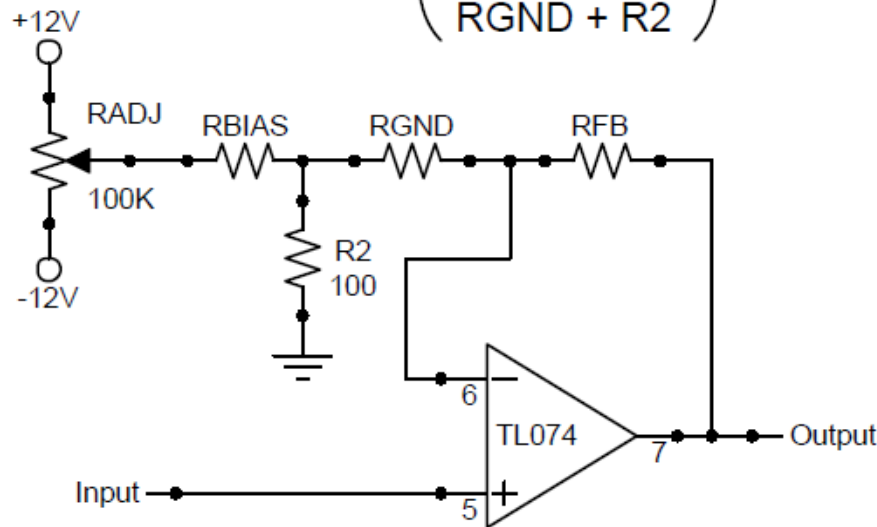
Input impedance determined as in examples above.

Offset Adjust or Biasing a Non-inverting Op Amp

For more information see: National Semiconductor LINEAR APPLICATIONS HANDBOOK
Linear Brief 9 - "Universal Balancing Techniques", Aug 1969 by Robert C. Dobkin

$$\text{Range} = \pm V \left(\frac{R2}{RBIAS} \right) \times \text{Gain}$$

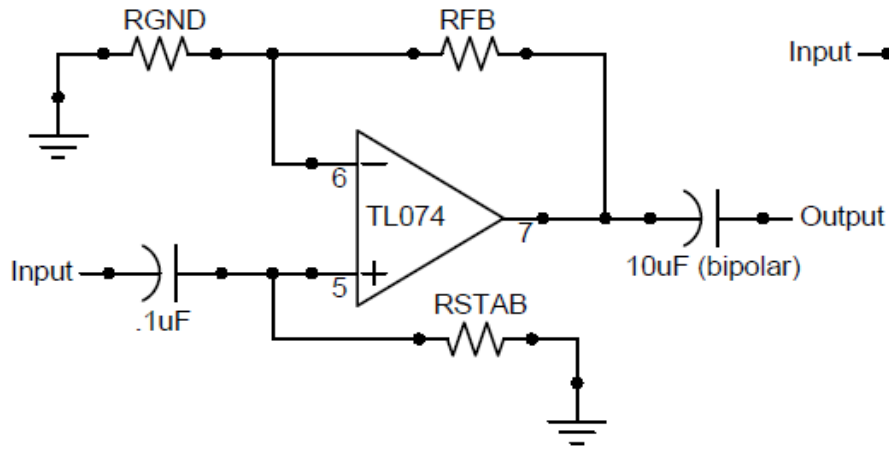
$$\text{Gain} = 1 + \left(\frac{RFB}{RGND + R2} \right)$$



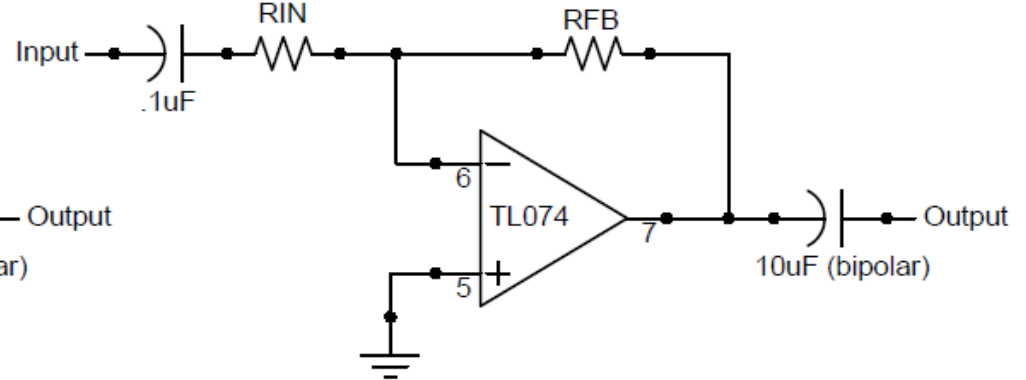
RBIAS should be selected based on the range of adjustment needed. Higher values of RBIAS will result in a smaller range of adjustment and vice versa. Again note that the offset voltage appears at the output multiplied by the op amp's gain.

Input/Output Coupling AC vs. DC

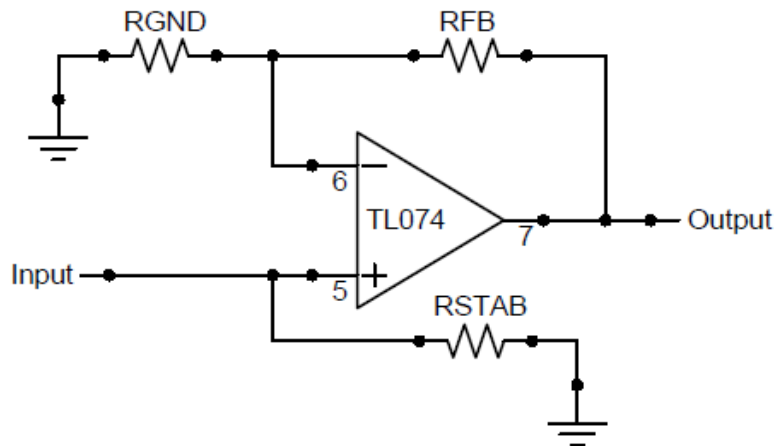
Non-inverting Configuration AC Coupled.



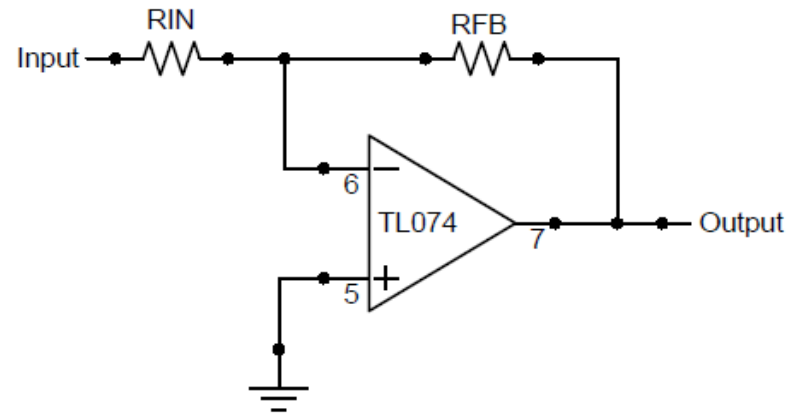
Inverting Configuration AC Coupled.



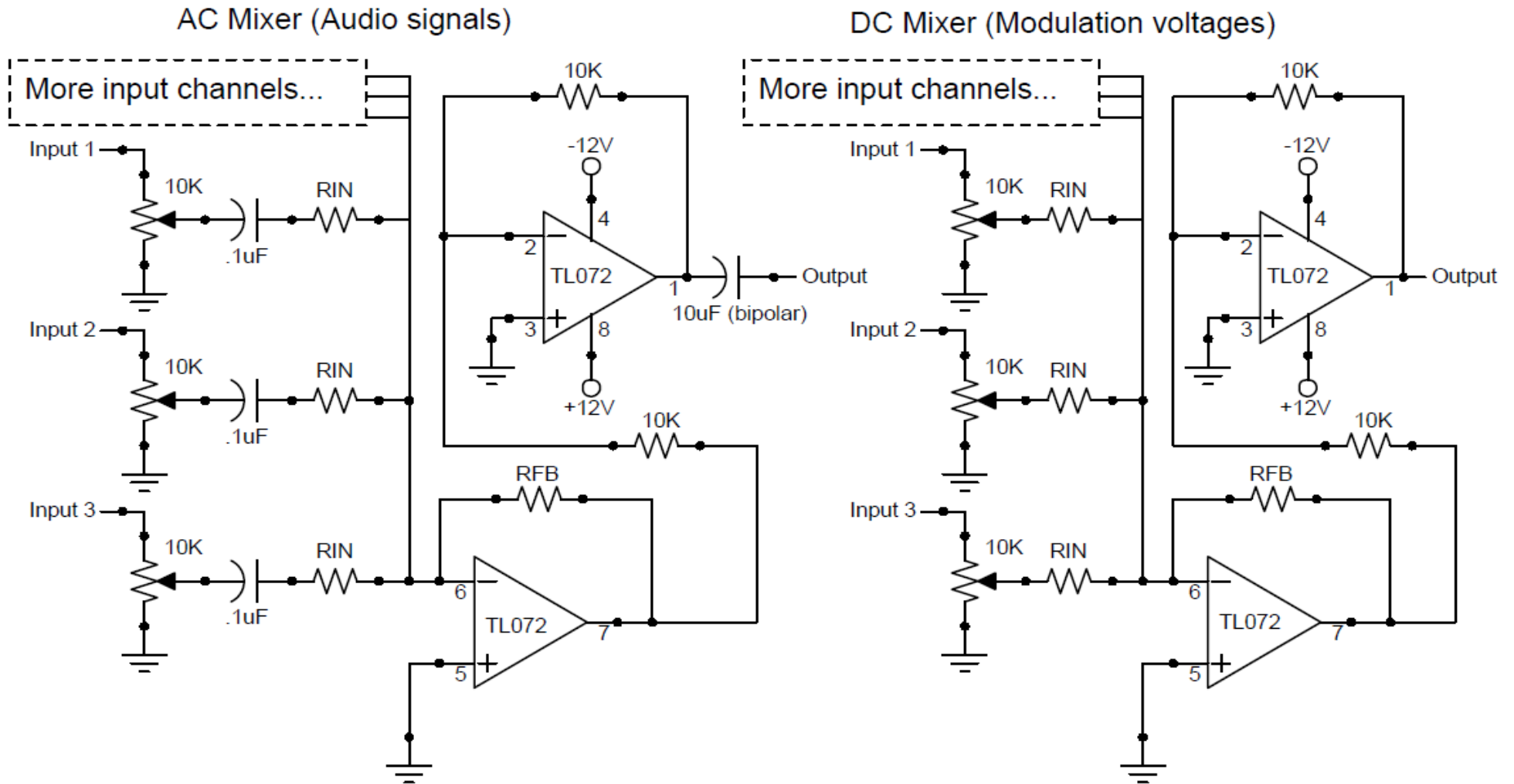
Non-inverting Configuration DC Coupled.



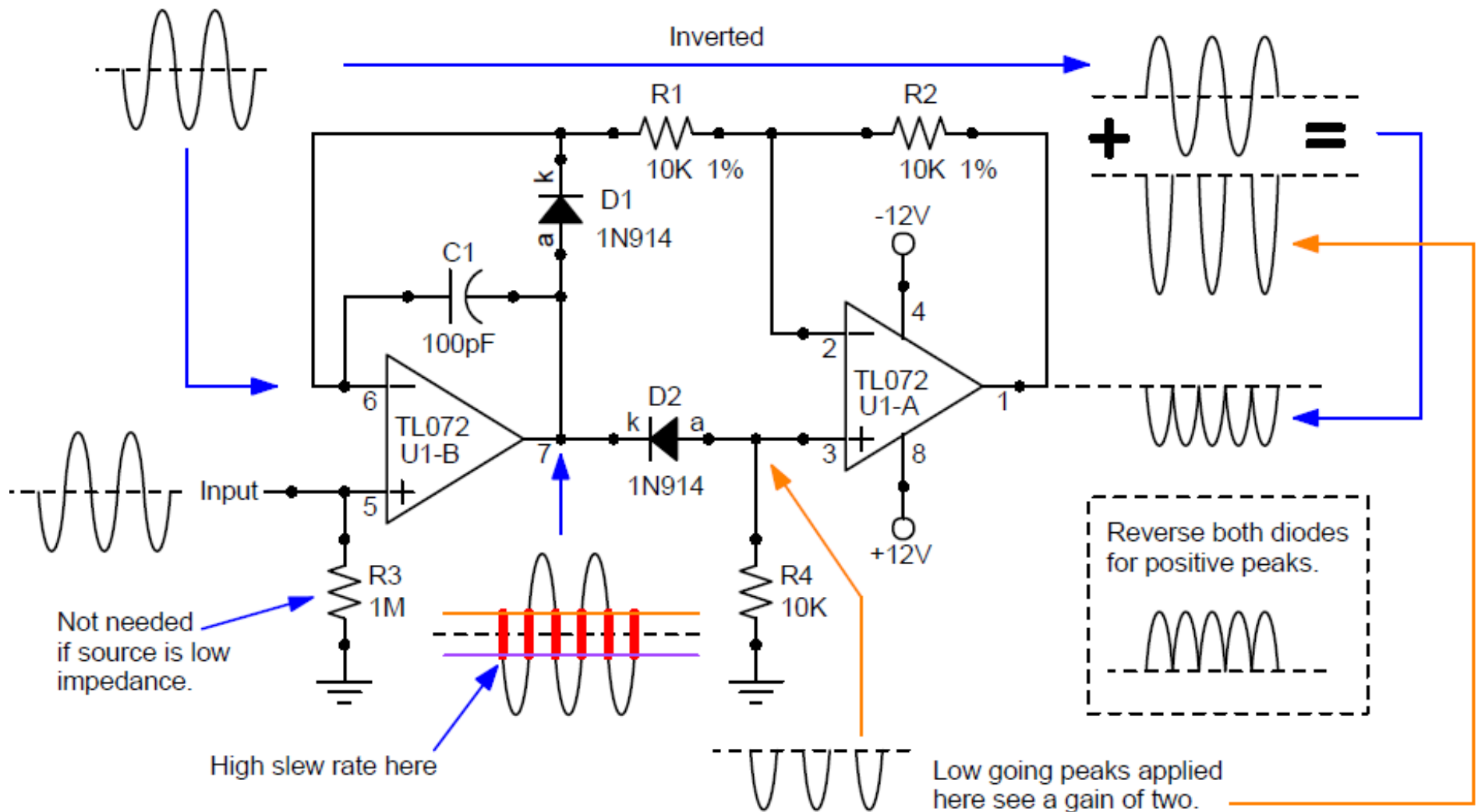
Inverting Configuration DC Coupled.



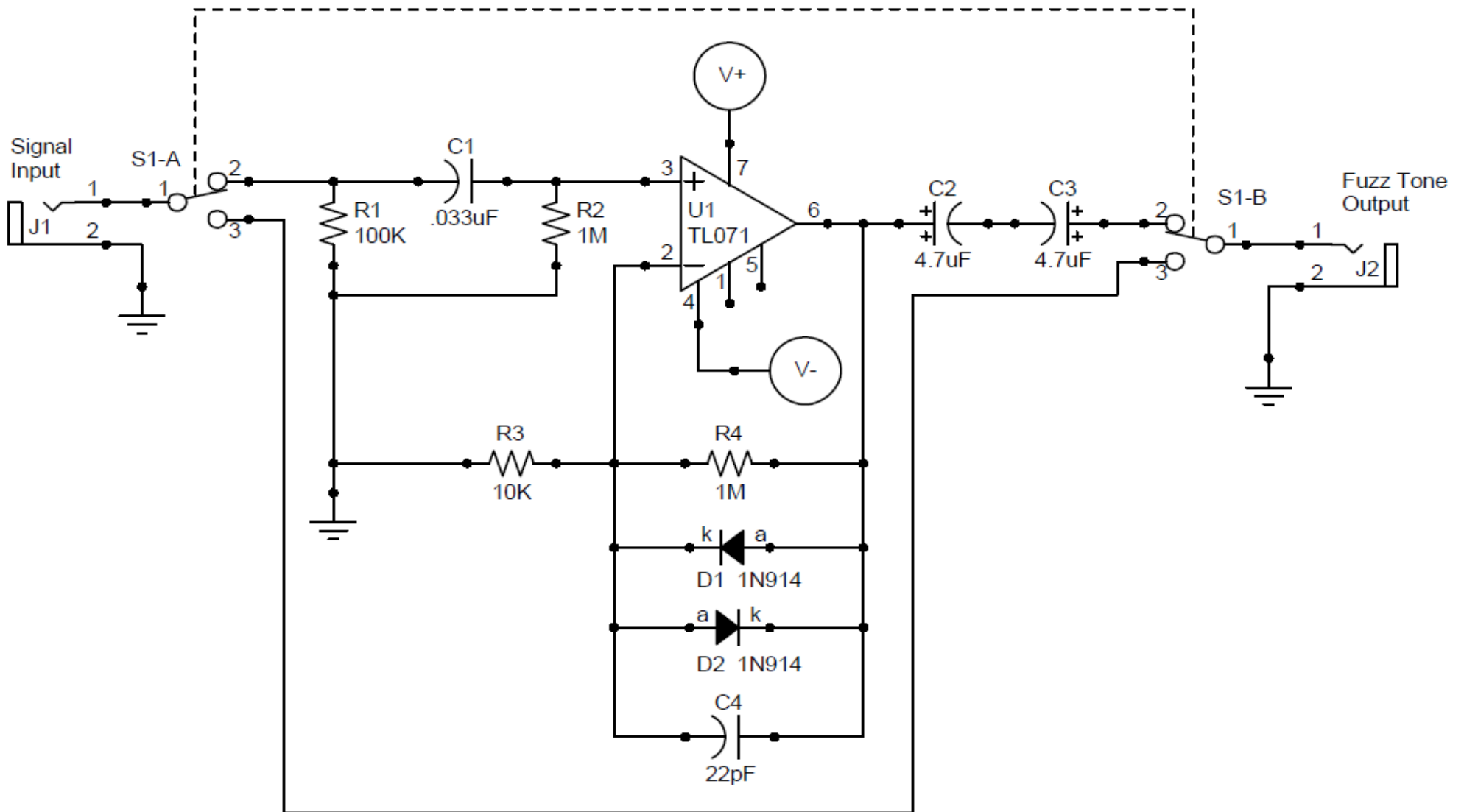
AC Mixer vs. DC Mixer



Active - Precision Full Wave Rectifier

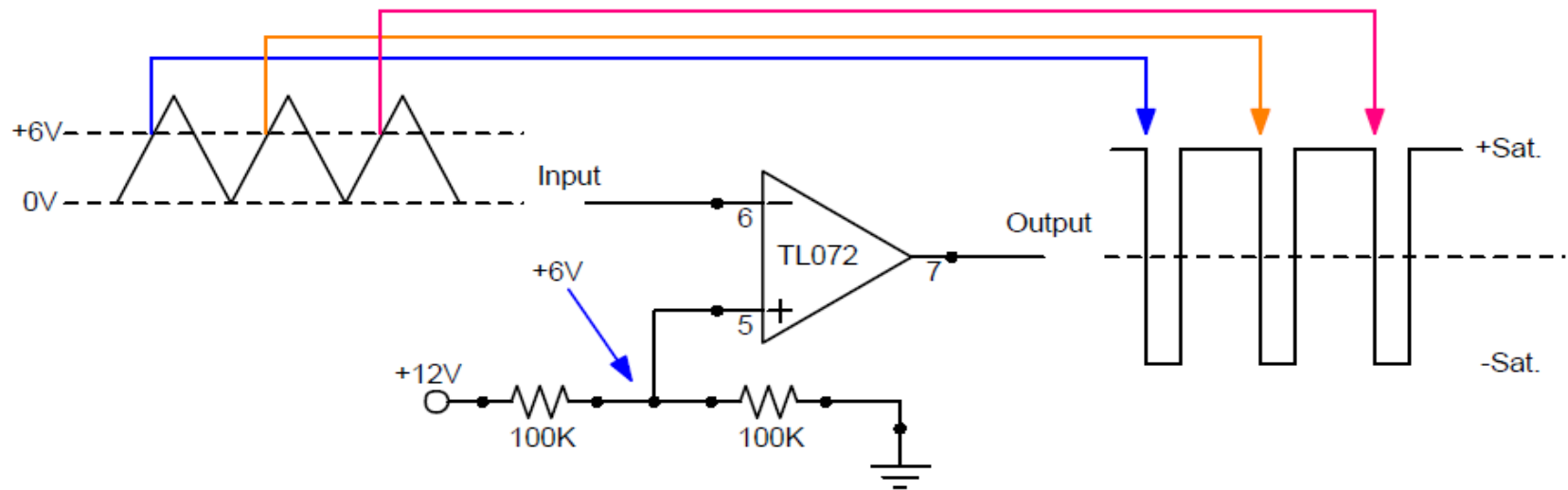


Fuzz Tone

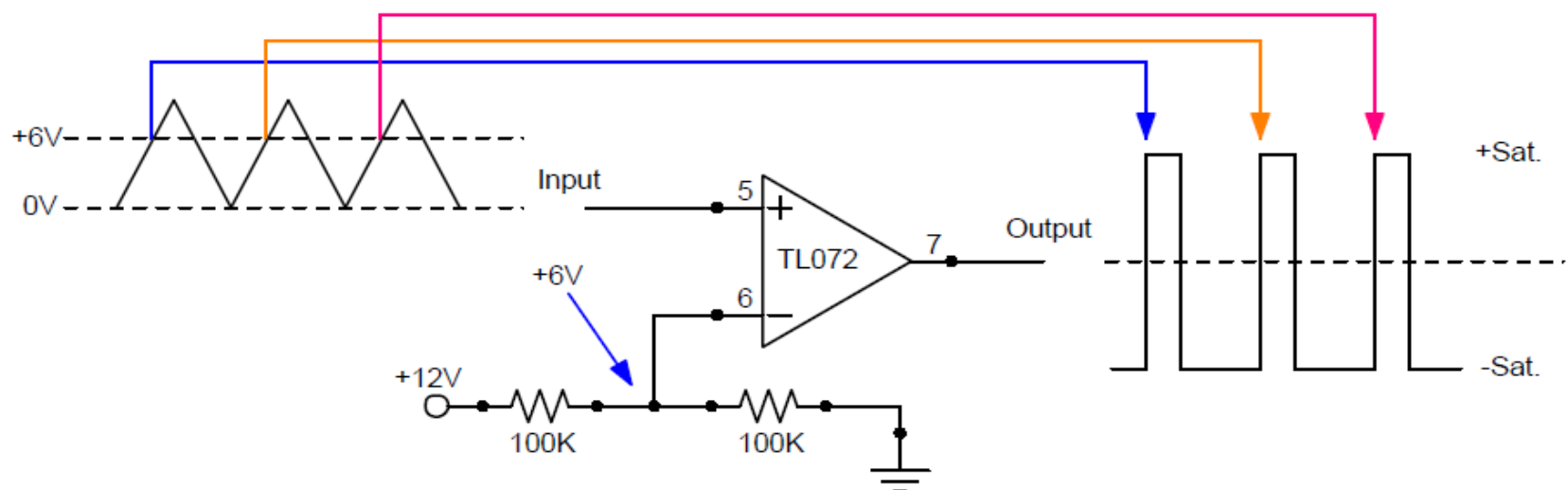


Comparators

Inverting Comparator

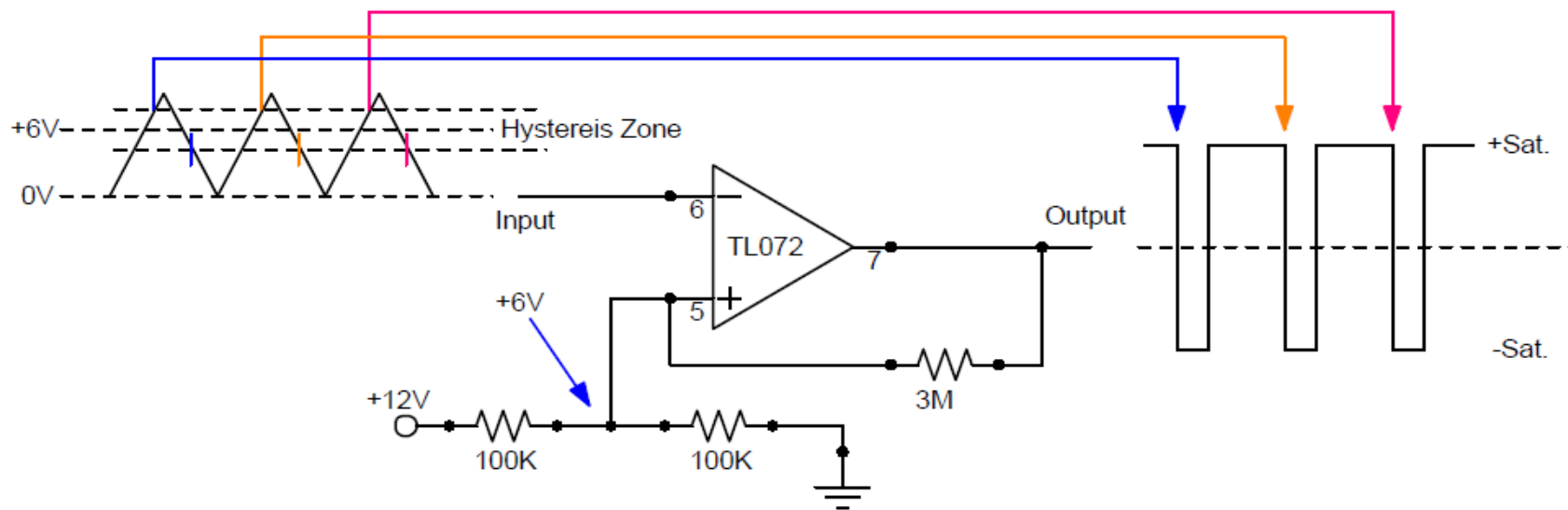


Non-inverting Comparator

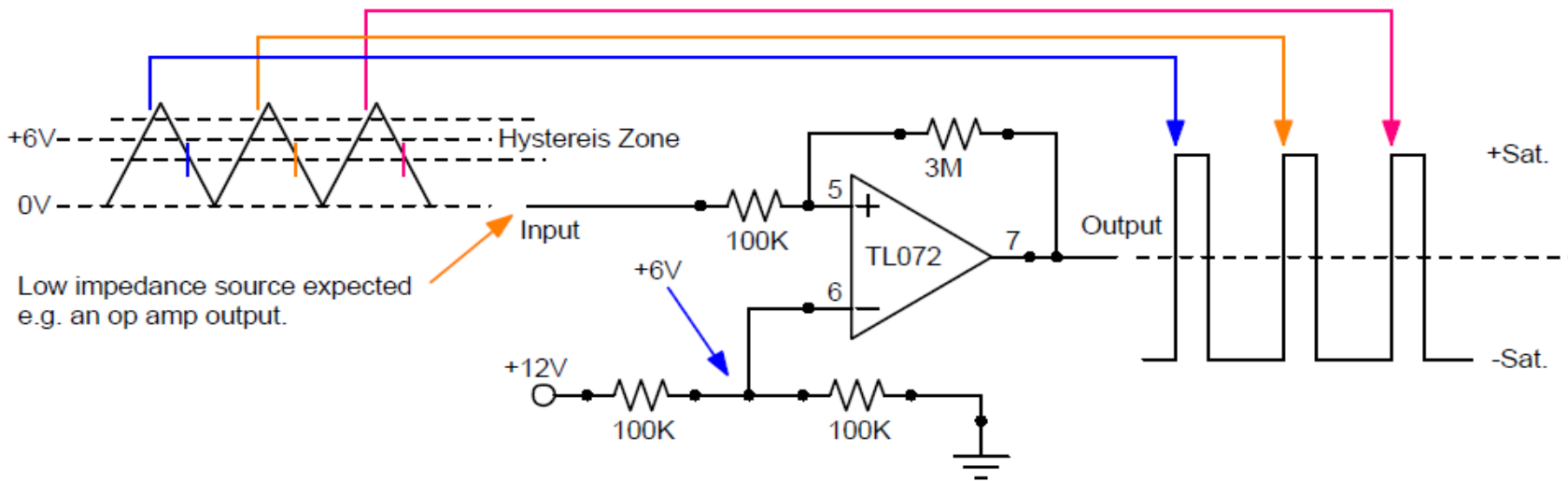


Comparator Hysteresis

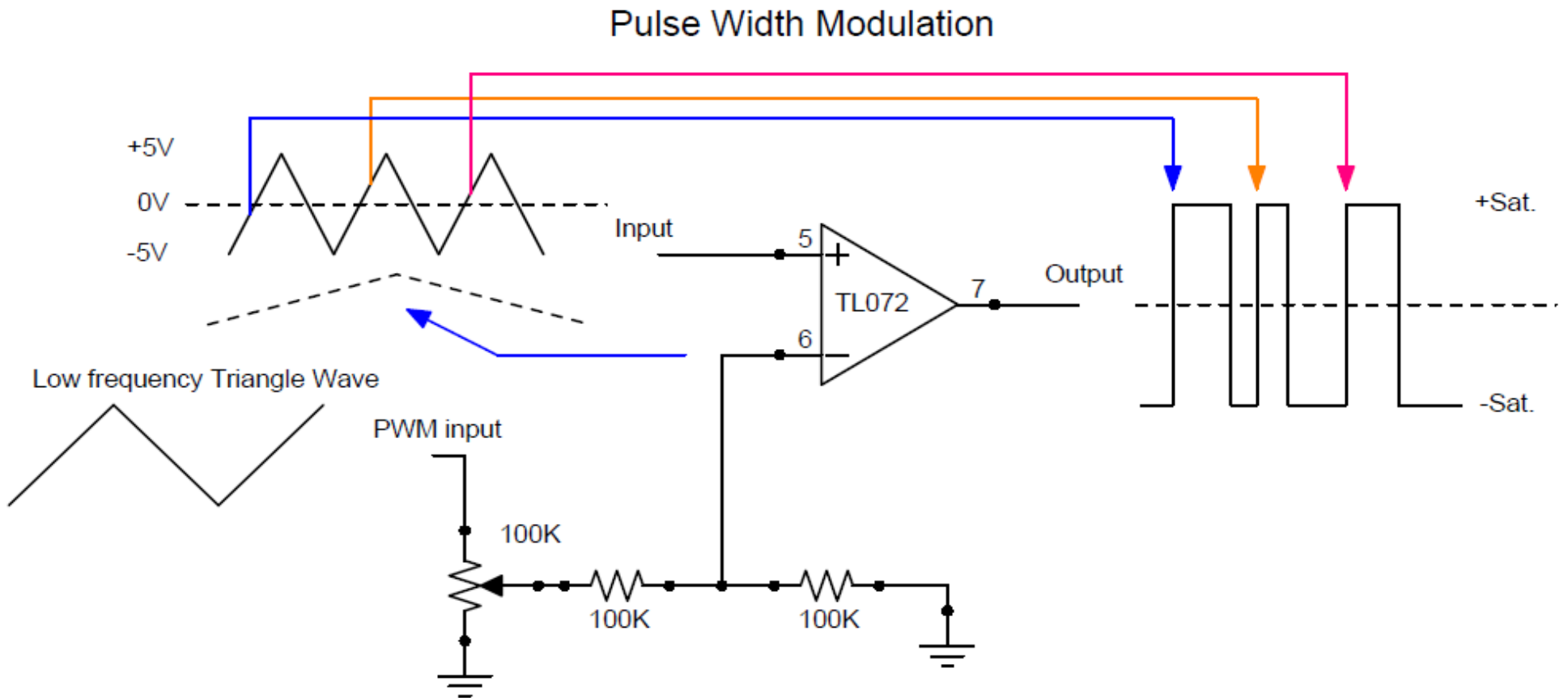
Inverting Comparator With Hysteresis



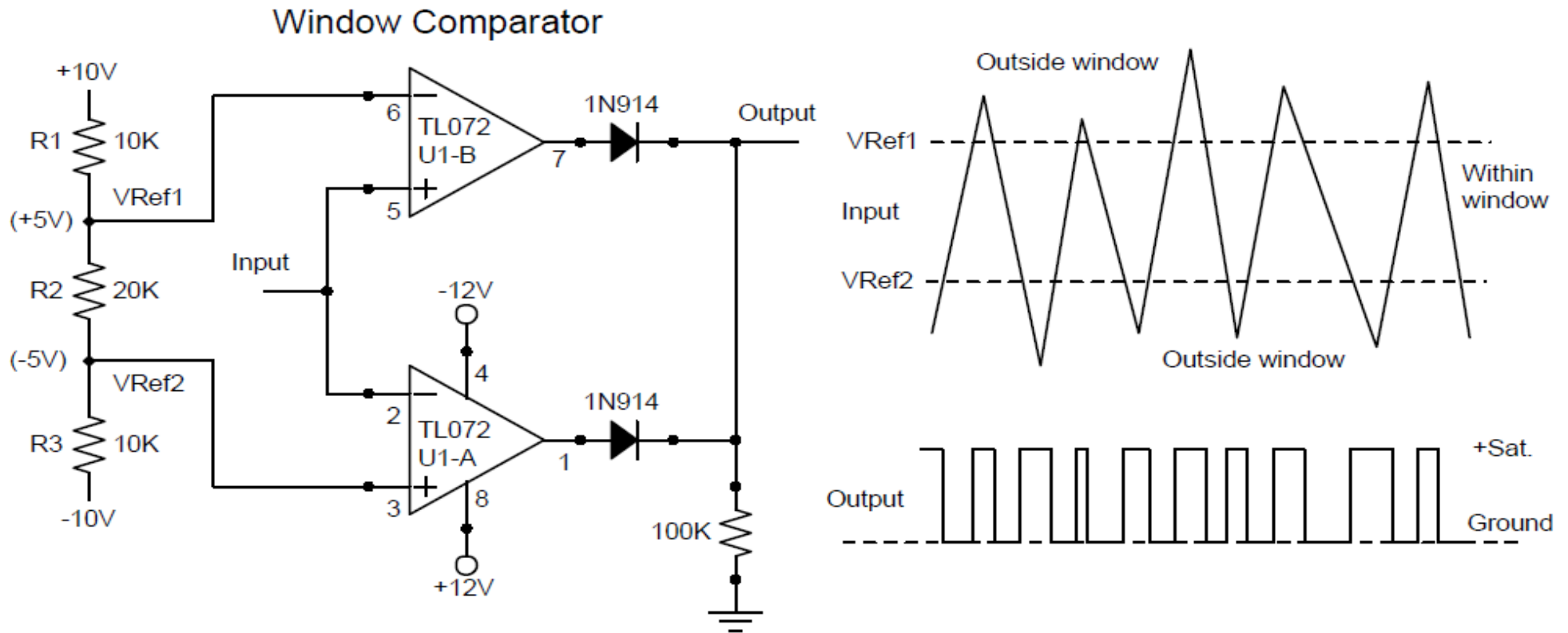
Non-inverting Comparator With Hysteresis



Pulse Width Modulation

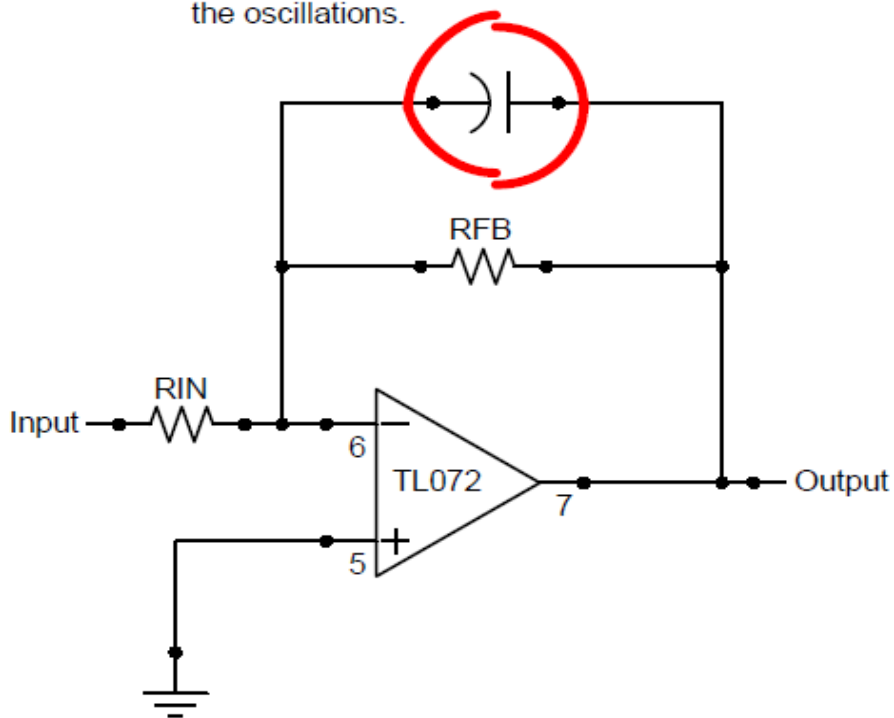


Window Comparator

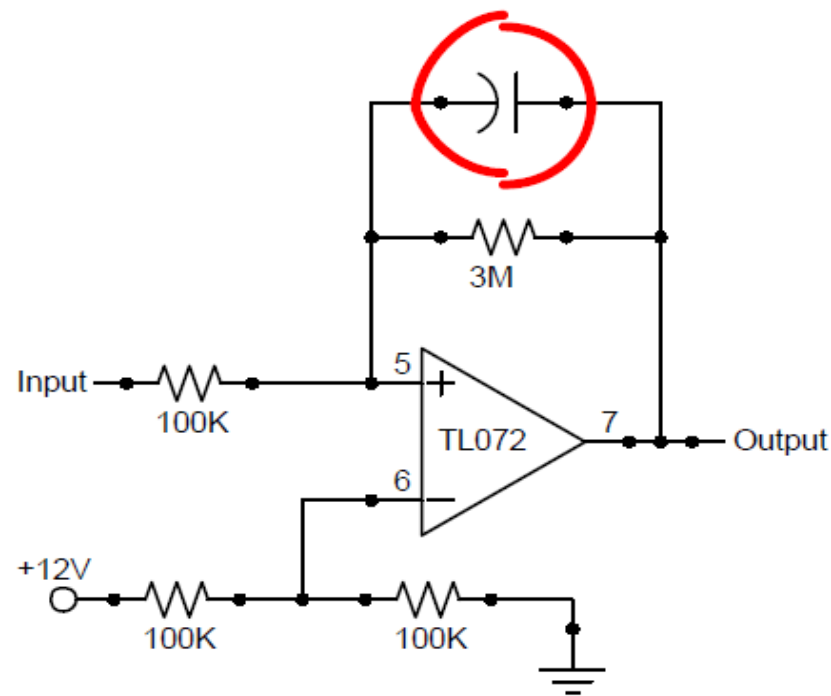


Capacitors to the Rescue

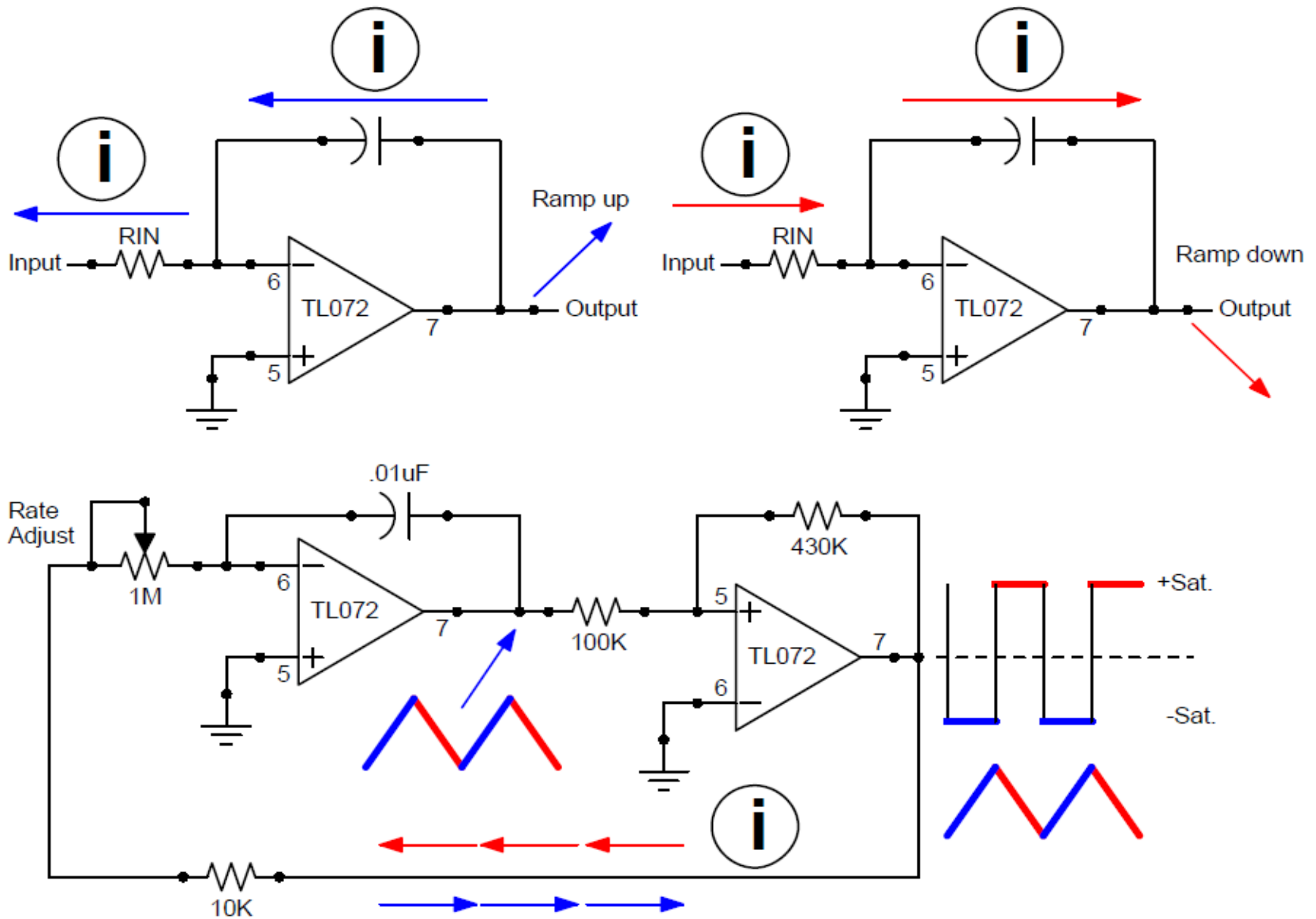
With high gain some op amps may oscillate at a high frequency. Placing a small value cap (10 to 100pF) across the negative feedback resistor can quell the oscillations.



To speed up your comparator's risetime place a small value cap (10 to 100pF) across the positive feedback resistor.

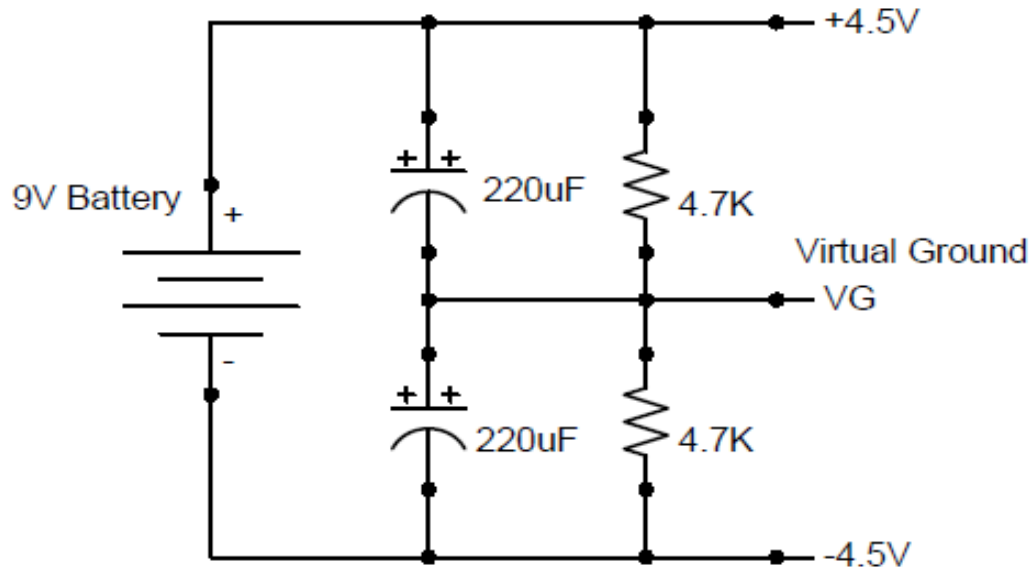


Integrators

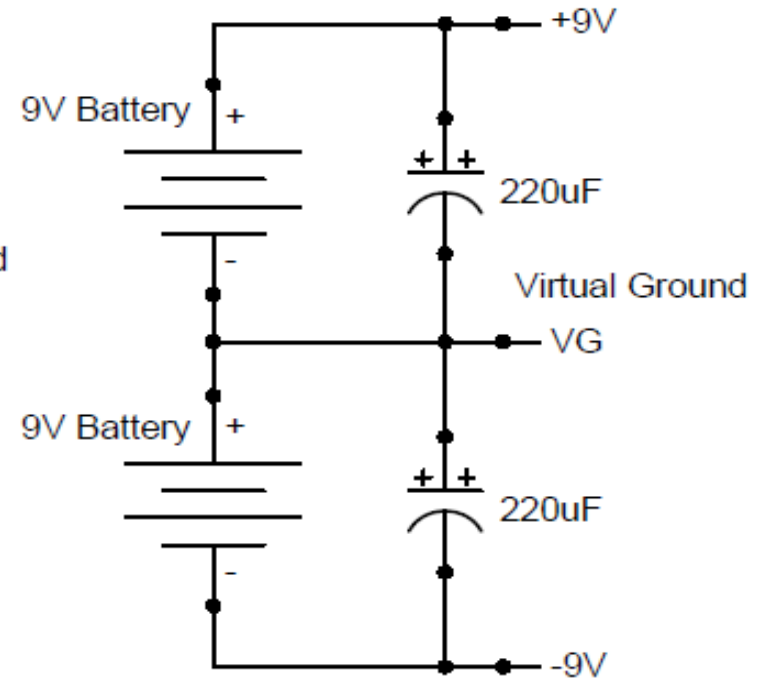


Battery Power

One nine volt battery can be used as a "split" $\pm 4.5V$ and virtual ground supply. Low current applications only.

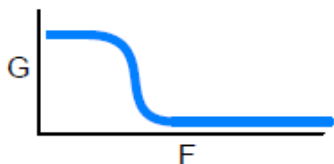
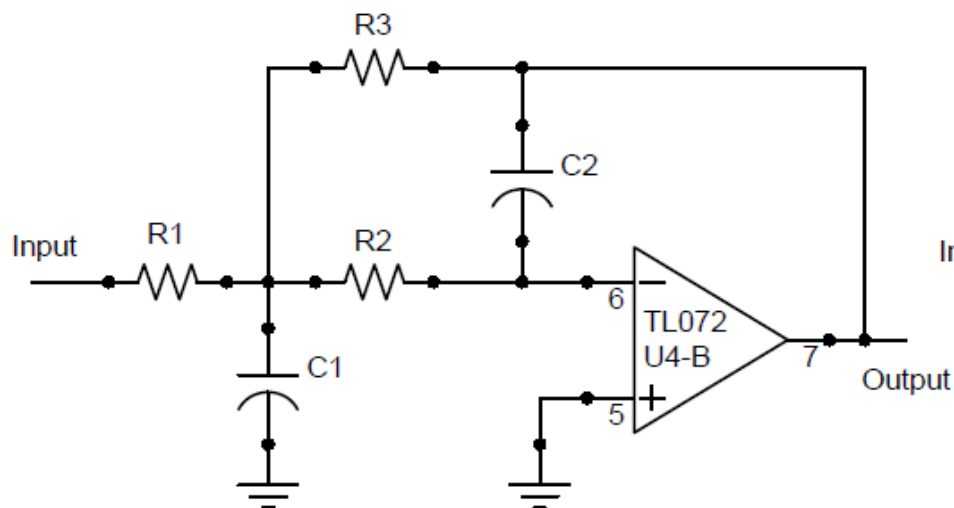


Two nine volt batteries can be used as a "split" $\pm 9V$ supply with a much lower impedance virtual ground. Low to medium current applications only.

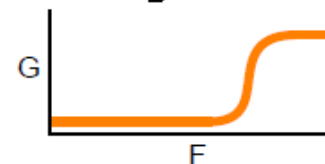
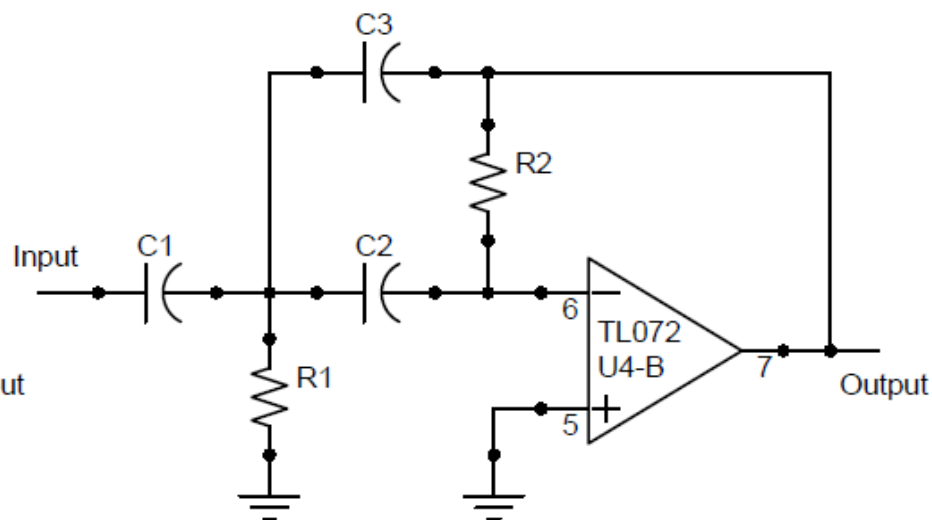


Active Lowpass, Highpass, and Bandpass Filters

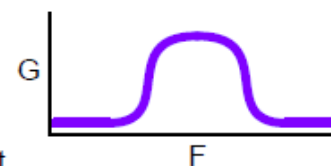
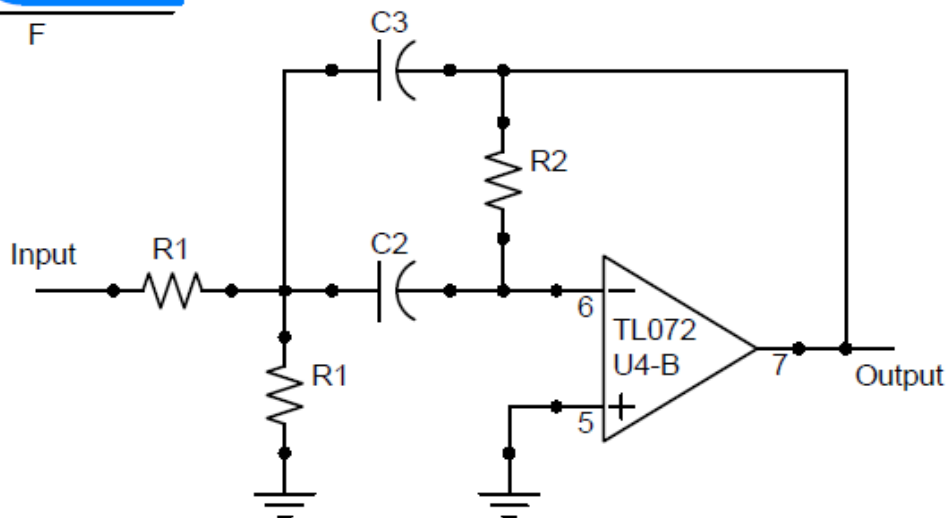
Multiple Feedback Lowpass Filter



Multiple Feedback Highpass Filter



Multiple Feedback Bandpass Filter



Online Filter Calculators You Should Know About

OKAWA Electric Design - Filter Design and Analysis

<http://sim.okawa-denshi.jp/en/Fkeisan.htm>

Texas Instruments FilterPro

<http://www.ti.com/tool/filterpro>

Texas Instruments WEBENCH Filter Designer

<http://www.ti.com/lscs/ti/analog/webench/webench-filters.page>